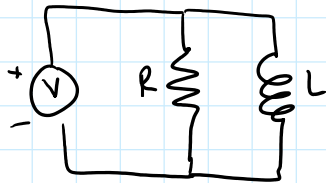
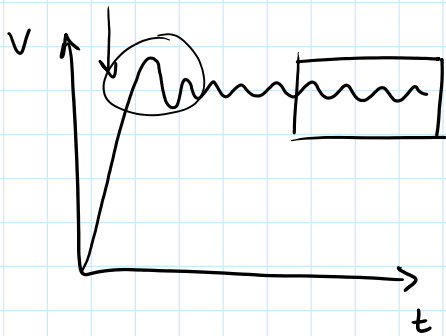


## 1. Phasors



→ Solve for  $V(t)$    
 ↗ Time Domain (Transient + Steady-state)   
 ↘ Phasor Domain (Steady-state)

Phasors → Amplitude, phase



Example for phasors:

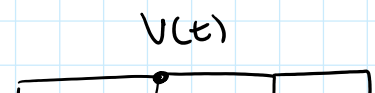
$$V_s(t) = \underbrace{A}_{\text{Amplitude}} \underbrace{\cos(\omega t + \phi)}_{\substack{\text{frequency} \quad \text{phase}}} \Rightarrow V_s(t) = \text{Re} \{ \underbrace{A e^{j(\omega t + \phi)}}_{\text{phasor}} \}$$

$$V_s(t) = \text{Re} \{ \underbrace{A e^{j(\omega t + \phi)}}_{\text{phasor}} \} = \text{Re} \{ \underbrace{A e^{j\phi}}_{\tilde{V}} e^{j\omega t} \}$$

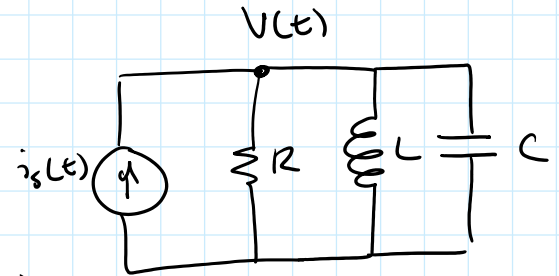
$$\tilde{V}_s \equiv \underbrace{A}_{\text{Amplitude}} e^{j \underbrace{\phi}_{\text{phase}}}, @ t = 0$$

$$V_s(t) = \text{Re} \{ \tilde{V}_s e^{j\omega t} \}$$

time domain version.



time domain version.



$$\Rightarrow i_s(t) = \frac{V(t)}{R} + i_L(t) + C \frac{dV(t)}{dt}$$

$$i_s(t) = \frac{V(t)}{R} + \frac{1}{L} \int V(t) dt + C \frac{dV(t)}{dt}$$

$$\frac{d}{dt} (\tilde{V}_s) = (j\omega) \tilde{V}_s$$

$$\int \tilde{V}_s dt = \frac{1}{j\omega} \tilde{V}$$

$$V_L(t) = L \frac{di_L(t)}{dt}$$

$$\frac{1}{L} \int V_L(t) dt$$

$$\tilde{I}_s = \frac{\tilde{V}}{R} + \frac{1}{j\omega L} \cdot \tilde{V} + (j\omega C) \tilde{V}$$

$$\tilde{V} = \frac{\tilde{I}_s}{\frac{1}{R} + \frac{1}{j\omega L} + j\omega C} = \frac{\tilde{I}_s}{\underbrace{\frac{1}{R}}_{\text{real}} + j \underbrace{\left(\omega C - \frac{1}{\omega L}\right)}_{\text{Imaginary}}}$$

$$\Rightarrow \text{Convert to time domain: } i_s(t) = A \cos\left(\omega t + \frac{\pi}{3}\right) A$$

$$i_s(t) = \text{Re} \{ \tilde{I}_s e^{j\omega t} \} A$$

$$\tilde{I}_s = A e^{j\pi/3}$$

$$\tilde{V} = \frac{\tilde{I}_s}{\left(\frac{1}{R} + j\left(\omega C - \frac{1}{\omega L}\right)\right)}$$

$$\text{Magnitude: } \sqrt{\left(\frac{1}{R}\right)^2 + \left(\omega C - \frac{1}{\omega L}\right)^2} = B$$

Phase

$$\tan \theta = \frac{\text{imaginary}}{\text{real}} = \frac{\omega C - 1/\omega L}{1/R}$$

$$\tan \theta = R(\omega C - 1/\omega L)$$

$$\theta = \tan^{-1}(R(\omega C - 1/\omega L))$$

$$v(t) = \operatorname{Re} \left\{ \frac{I_0 e^{j\pi/3} e^{j\omega t}}{B e^{j\theta}} \right\}$$

$$v(t) = \operatorname{Re} \left\{ \frac{I_0}{B} e^{j\omega t} e^{j(\pi/3 - \theta)} \right\}$$

$$v(t) = \frac{I_0}{\sqrt{(1/R)^2 + (\omega C - 1/\omega L)^2}} \cos(\omega t + \frac{\pi}{3} - \theta)$$

$$\theta = \tan^{-1}(R(\omega C - 1/\omega L))$$

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Characteristic Impedance:

⇒ Example - Reflection of series RC Load

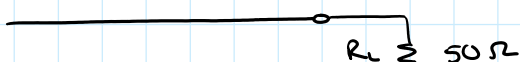
Consider a  $100 \Omega$  transmission line,  $Z_0 = 100 \Omega$

Connected to load: resistor  $50 \Omega$  in series with  $10 \text{ pF}$  capacitor:

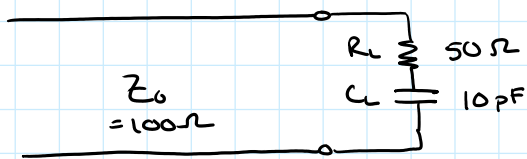
$$\rightarrow R_L = 50 \Omega$$

$$C_L = 10 \text{ pF} = 10^{-11} \text{ F}$$

$$f = 100 \text{ MHz} = 10^8 \text{ Hz}$$



n. Find the reflection



T = 100 MHz

Q: Find the reflection coefficient for 100 MHz signal.

Knowns:  $Z_0, R_L, C_L, f$

$$\Gamma = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{Z_L/Z_0 - 1}{Z_L/Z_0 + 1} = \frac{Z_L - 1}{Z_L + 1}, \quad Z_L = \frac{Z_L}{Z_0}$$

$$Z_L = \frac{Z_L}{Z_0}, \quad Z_L = R_L - j/\omega C_L$$

$$Z_L = \frac{R_L - j/\omega C_L}{Z_0} = \frac{1}{100} \left( 50 - j \frac{1}{2\pi \times 10^8 \times 10^{-11}} \right)$$

$$Z_L = 0.5 - j1.59 \Omega$$

Solving reflection

$$\Gamma = \frac{Z_L - 1}{Z_L + 1} = \frac{(0.5 - j1.59) - 1}{(0.5 - j1.59) + 1} = \frac{-0.5 - j1.59}{1.5 - j1.59}$$

$$\Gamma = \frac{-0.5 - j1.59}{1.5 - j1.59} \rightarrow A = \sqrt{(0.5)^2 + (1.59)^2} \Rightarrow -A$$

$$= 1.67$$

$$\phi_A = \tan^{-1} \left( \frac{1.59}{0.5} \right)$$

$$= 72.5^\circ$$

$$-A e^{j\phi_A}$$

$$B = \sqrt{(1.5)^2 + (1.59)^2}$$

$$= 2.19$$

$$\phi_R = \tan^{-1} \left( \frac{-1.59}{-0.5} \right) = -46.7^\circ$$

$$\phi_B = \tan^{-1}\left(\frac{-1.59}{1.5}\right) = -46.7^\circ$$

$$-Ae^{j\omega t}$$

$$Be^{j\phi_B}$$

$$\Gamma = \frac{-1.67e^{j72.6^\circ}}{2.19e^{-j46.7^\circ}} = -0.76e^{j119.3^\circ}$$

$$\Gamma = e^{-j180^\circ} e^{j119.3^\circ} (0.76)$$

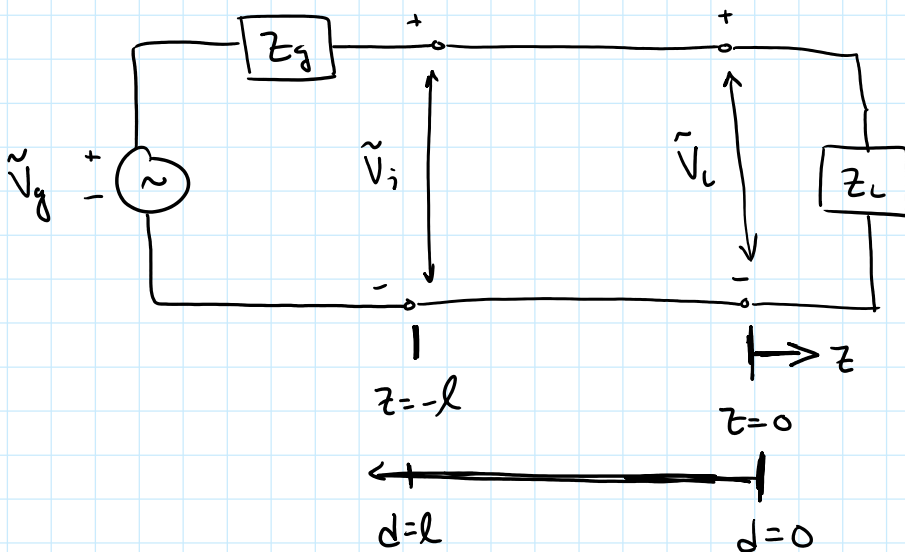
$$\Gamma = 0.76e^{-j60.7^\circ} = 0.76 \angle -60.7^\circ$$

$$\frac{V_o^-}{V_o^+} = \Gamma = \frac{Z_L - Z_0}{Z_L + Z_0}$$



Standing Waves:

$$Z_0 = \sqrt{\frac{R' + j\omega L'}{G' + j\omega C'}}$$



$$\star \quad |\tilde{V}(d)| = |V_o^+| \left[ 1 + |\Gamma|^2 + 2|\Gamma| \cos(2\beta d - \theta_r) \right]^{1/2}$$

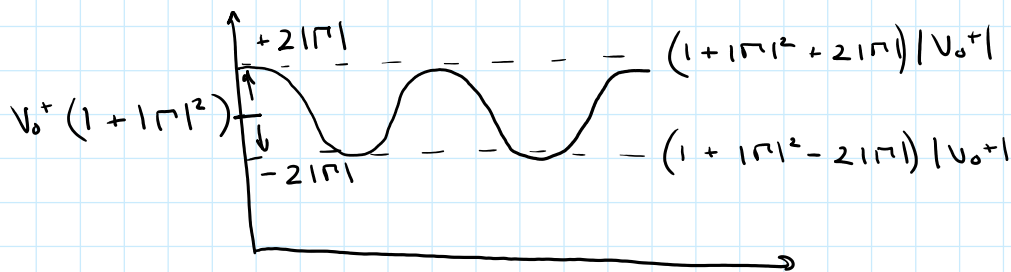
$$\star \quad |\tilde{V}(d)| = |V_o^+| \left[ 1 + |\Gamma|^2 + 2|\Gamma| \cos(2\beta d - \theta_r) \right]^{1/2}$$

$$\star \quad |\tilde{I}(d)| = \frac{|V_o^+|}{Z_o} \left[ 1 + |\Gamma|^2 - 2|\Gamma| \cos(2\beta d - \theta_r) \right]^{1/2}$$

→ Maximum:  $|\tilde{V}(d)|$  max occurs when:  $\underline{\underline{2\beta d - \theta_r = 2n\pi}}$

→ Minimum  $|\tilde{V}(d)|$  min occurs when:  $2\beta d - \theta_r = (2n+1)\pi$

$$\left[ \underbrace{1 + |\Gamma|^2}_{\text{DC offset}} + \underbrace{2|\Gamma| \cos(2\beta d - \theta_r)}_{\text{oscillation}} \right]^{1/2}$$



Note: spatial  $(\lambda)$  of a standing wave =  $\lambda/2$

$$|\tilde{V}|_{\max} = |V_o^+| (1 + |\Gamma|)$$

$$\Rightarrow 2\beta d_{\max} - \theta_r = 2\pi n$$

$$\left[ d_{\max} = \frac{\theta_r + 2n\pi}{2\beta} = \frac{\theta_r \lambda}{4\pi} + \frac{\lambda}{2} \right], \quad \beta = \frac{2\pi}{\lambda}$$

$$\begin{cases} n = 1, 2, \dots & \text{if } \theta_r < 0 \\ n = 0, 1, 2, \dots & \text{if } \theta_r \geq 0 \end{cases} \quad -\pi \leq \theta_r \leq \pi$$

$$\Theta_r \geq 0 : \quad d_{\max} = \frac{\Theta_r \lambda}{4\pi} \quad \Rightarrow \quad n=0$$

$$\Theta_r < 0 : \quad d_{\max} = \frac{\Theta_r \lambda}{4\pi} + \frac{\lambda}{2} \quad \left. \vphantom{\frac{\Theta_r \lambda}{4\pi}} \right\} \text{ this occurs at } n=1 \text{ first}$$

$$|\tilde{V}_{\min}| = |V_s^+| (1 - |\Gamma|)$$

$$d_{\min} \begin{cases} d_{\max} + \frac{\lambda}{4} & \text{if } d_{\max} < \frac{\lambda}{4} \\ d_{\max} - \frac{\lambda}{4} & \text{if } d_{\max} \geq \frac{\lambda}{4} \end{cases}$$